

Lecture #10

ANNOUNCEMENT

- Quiz #2 next Thursday (2/27) covering
 - carrier action (drift, diffusion, R-G)
 - continuity & minority-carrier diffusion equations
 - metal-semiconductor contacts

OUTLINE

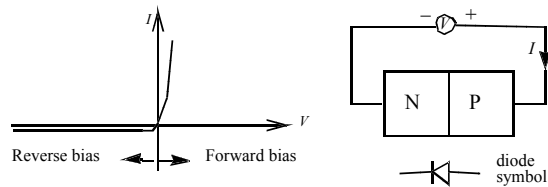
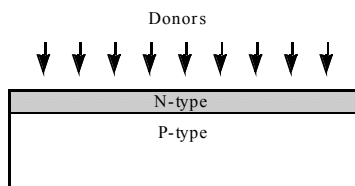
- pn junction electrostatics

Reading: Chapter 5

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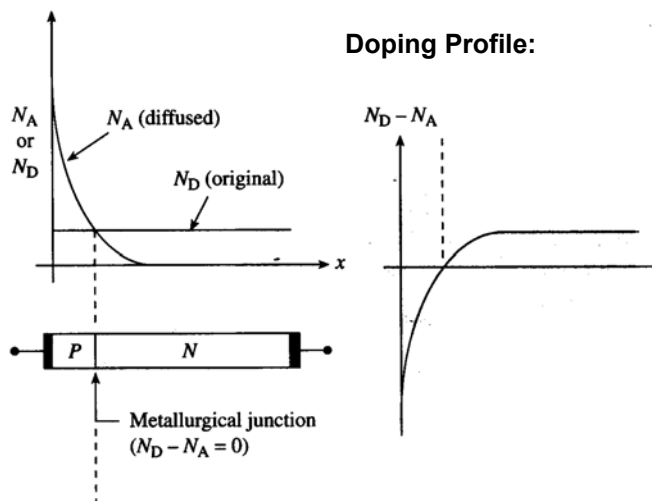
pn Junctions



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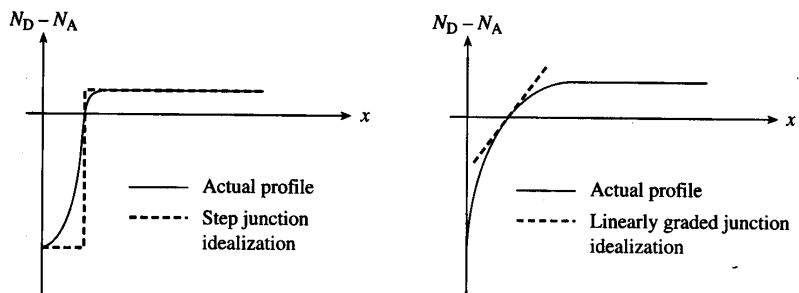
Terminology



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Idealized Junctions



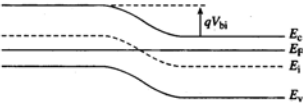
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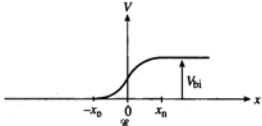
Energy Band Diagram

Qualitative Electrostatics

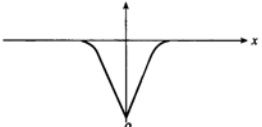
Band diagram



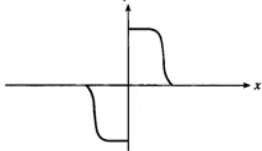
Electrostatic potential



Electric field



Charge density



“Game Plan” for Obtaining $\rho(x)$, $\mathcal{E}(x)$, $V(x)$

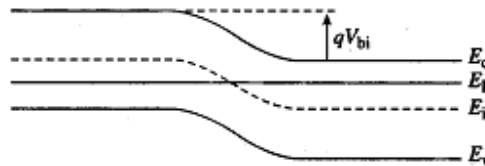
- Find the built-in potential V_{bi}
- Use the depletion approximation $\rightarrow \rho(x)$
(depletion-layer widths x_p , x_n unknown)
- Integrate $\rho(x)$ to find $\mathcal{E}(x)$
 - boundary conditions $\mathcal{E}(-x_p)=0$, $\mathcal{E}(x_n)=0$
- Integrate $\mathcal{E}(x)$ to obtain $V(x)$
 - boundary conditions $V(-x_p)=0$, $V(x_n)=V_{bi}$
- For $\mathcal{E}(x)$ to be continuous at $x=0$, $N_A x_p = N_D x_n$
 \rightarrow solve for x_p , x_n

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Built-In Potential V_{bi}

$$qV_{bi} = \Phi_{\text{Sp-side}} - \Phi_{\text{Sn-side}} = (E_i - E_F)_{\text{p-side}} + (E_F - E_i)_{\text{n-side}}$$



For non-degenerately doped material:

$$\begin{aligned} (E_i - E_F)_{\text{p-side}} &= kT \ln\left(\frac{p}{n_i}\right) & (E_F - E_i)_{\text{n-side}} &= kT \ln\left(\frac{n}{n_i}\right) \\ &= kT \ln\left(\frac{N_A}{n_i}\right) & &= kT \ln\left(\frac{N_D}{n_i}\right) \end{aligned}$$

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V_{bi} for “One-Sided” pn Junctions

$$qV_{bi} = (E_i - E_F)_{p\text{-side}} + (E_F - E_i)_{n\text{-side}}$$

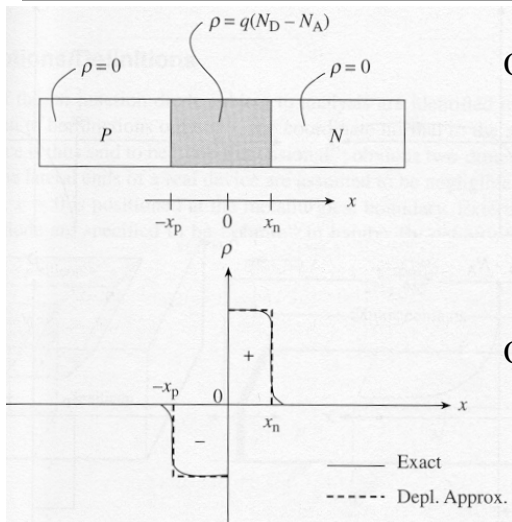
p⁺n junction

n⁺p junction

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The Depletion Approximation



On the *p*-side, $\rho = -qN_A$

$$\frac{d\mathcal{E}}{dx} = -\frac{qN_A}{\epsilon_s}$$

$$\mathcal{E}(x) = -\frac{qN_A}{\epsilon_s}x + C_1 = -\frac{qN_A}{\epsilon_s}(x + x_p)$$

On the *n*-side, $\rho = qN_D$

$$\mathcal{E}(x) = \frac{-qN_D}{\epsilon_s}(x_n - x)$$

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Electric Field in the Depletion Layer

The electric field is continuous at $x = 0$

$$\Rightarrow N_A x_p = N_D x_n$$

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Electrostatic Potential in the Depletion Layer

On the *p-side*:

$$V(x) = \frac{qN_A}{2\epsilon_s} (x + x_p)^2 + D_1$$

(arbitrarily choose the voltage at $x = x_p$ to be 0)

On the *n-side*:

$$V(x) = -\frac{qN_D}{2\epsilon_s} (x_n - x)^2 + D_2 = V_{bi} - \frac{qN_D}{2\epsilon_s} (x_n - x)^2$$

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-
- At $x = 0$, expressions for p-side and n-side must be equal:

- We also know that $N_A x_p = N_D x_n$

Depletion Layer Width

- Eliminating x_p , we have:

$$x_n = \sqrt{\frac{2\epsilon_s V_{bi}}{q} \left(\frac{N_A}{N_D(N_A + N_D)} \right)}$$

- Eliminating x_n , we have:

$$x_p = \sqrt{\frac{2\epsilon_s V_{bi}}{q} \left(\frac{N_D}{N_A(N_A + N_D)} \right)}$$

- Summing, we have:

$$x_n + x_p = W = \sqrt{\frac{2\epsilon_s V_{bi}}{q} \left(\frac{1}{N_A} + \frac{1}{N_D} \right)}$$

One-Sided Junctions

If $N_A \gg N_D$ as in a p⁺n junction:

$$W = \sqrt{\frac{2\epsilon_s V_{bi}}{qN_D}} \approx x_n$$

$$x_p = x_n N_D / N_A \cong 0$$

What about a n⁺p junction?

$$W = \sqrt{2\epsilon_s V_{bi} / qN} \quad \text{where} \quad \frac{1}{N} = \frac{1}{N_D} + \frac{1}{N_A} \approx \frac{1}{\text{lighter dopant density}}$$

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Peak Electric Field

$$\left| \int \mathcal{E} \, dx \right| = \frac{1}{2} |\mathcal{E}(0)| W = V_{bi} - V_A$$

- For a one-sided junction, $W \cong \sqrt{\frac{2\epsilon_s}{qN} (V_{bi} - V_A)}$

$$\text{so } \mathcal{E}(0) = \frac{2(V_{bi} - V_A)}{W} \cong \sqrt{\frac{2qN(V_{bi} - V_A)}{\epsilon_s}}$$

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Example

A p^+n junction has $N_A = 10^{20} \text{ cm}^{-3}$ and $N_D = 10^{17} \text{ cm}^{-3}$. What is a) its built-in potential, b) W , c) x_n , and d) x_p ?

Solution:

$$a) \quad V_{bi} = \frac{E_G}{2q} + \frac{kT}{q} \ln \frac{N_D}{n_i} \approx 1 \text{ V}$$

$$b) \quad W \approx \sqrt{\frac{2\epsilon_s V_{bi}}{qN_D}} = \left(\frac{2 \times 12 \times 8.85 \times 10^{-14} \times 1}{1.6 \times 10^{-19} \times 10^{17}} \right)^{1/2} = 0.12 \text{ } \mu\text{m}$$

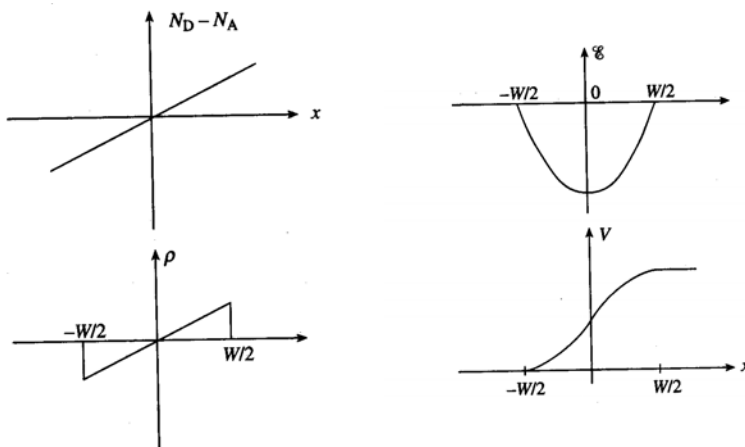
$$c) \quad x_n \approx W = 0.12 \text{ } \mu\text{m}$$

$$d) \quad x_p = x_n N_D / N_A = 1.2 \times 10^{-4} \text{ } \mu\text{m} = 1.2 \text{ } \text{\AA} \approx 0$$

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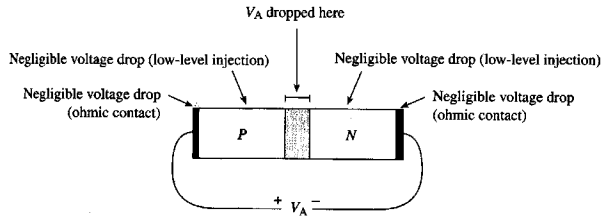
Linearly Graded Junction



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Biased PN Junctions

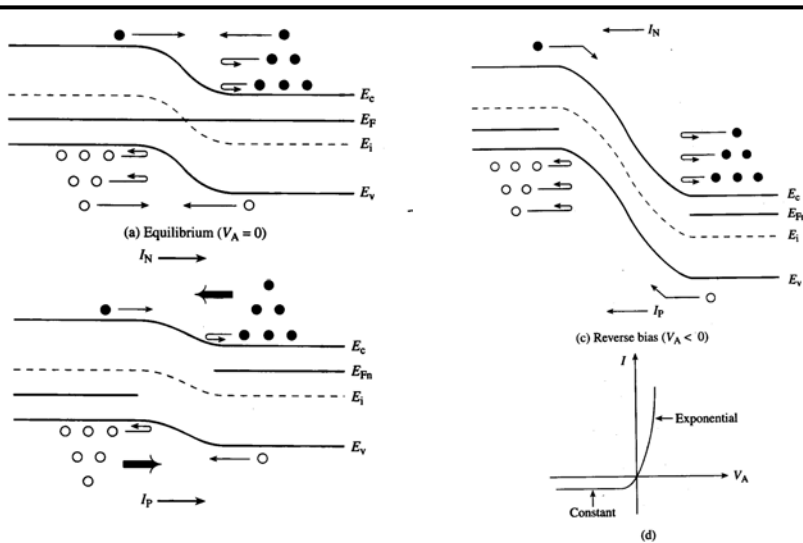


REQUIREMENT: $V_A < V_{bi}$, otherwise, we cannot assume low-level injection

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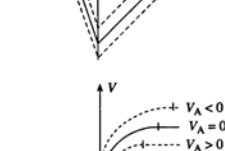
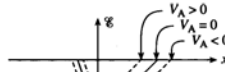
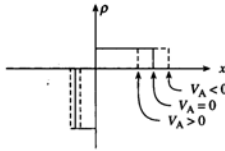
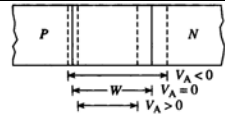
Current Flow - Qualitative



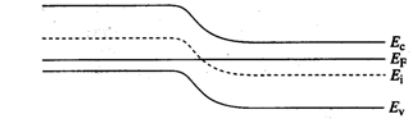
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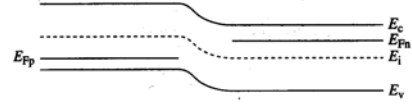
Effect of Bias on Electrostatics



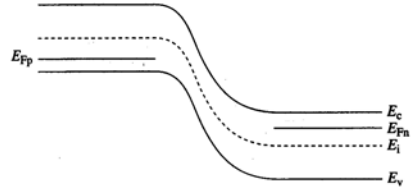
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(a) Equilibrium ($V_A = 0$)



(b) Forward bias ($V_A > 0$)



(c) Reverse bias ($V_A < 0$)